

Colpitts oscillator

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A **Colpitts oscillator**, invented in 1918 by American engineer Edwin H. Colpitts,^[1] is one of a number of designs for LC oscillators, electronic oscillators that use a combination of inductors (L) and capacitors (C) to produce an oscillation at a certain frequency. The distinguishing feature of the Colpitts oscillator is that the feedback for the active device is taken from a voltage divider made of two capacitors in series across the inductor.^{[2][3][4][5]}

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Overview

The Colpitts circuit, like other LC oscillators, consists of a gain device (such as a bipolar junction transistor, field effect transistor, operational amplifier, or vacuum tube) with its output connected to its input in a feedback loop containing a parallel LC circuit (tuned circuit) which functions as a bandpass filter to set the frequency of oscillation.

A Colpitts oscillator is the electrical dual of a Hartley oscillator, where the feedback signal is taken from an "inductive" voltage divider consisting of two coils in series (or a tapped coil). Fig. 1 shows the common-base Colpitts circuit. L and the series combination of C_1 and C_2 form the parallel resonant tank circuit which determines the frequency of the oscillator. The voltage across C_2 is applied to the base-emitter junction of the transistor, as feedback to create oscillations. Fig. 2 shows the common-collector version. Here the voltage across C_1 provides feedback. The frequency of oscillation is approximately the resonant frequency of the LC circuit, which is the series combination of the two capacitors in parallel with the inductor

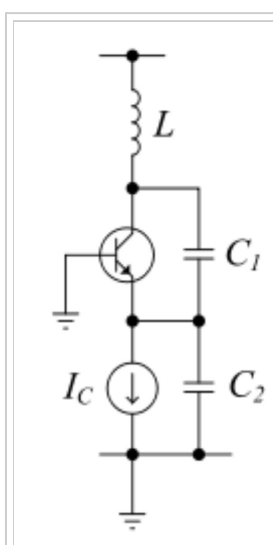


Figure 1: Simple common base Colpitts oscillator (with simplified biasing)

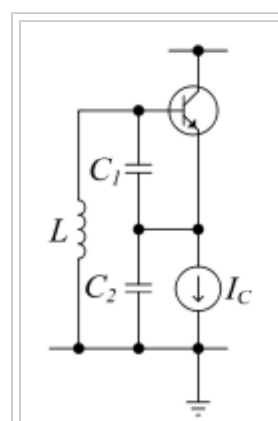


Figure 2: Simple common collector Colpitts oscillator (with simplified biasing)

$$f_0 = \frac{1}{2\pi\sqrt{L\left(\frac{C_1C_2}{C_1+C_2}\right)}}$$

The actual frequency of oscillation will be slightly lower due to junction capacitances and resistive loading of the transistor.

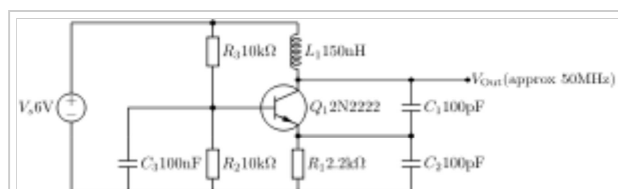


Figure 3: Practical common base Colpitts oscillator (with an oscillation frequency of ~50 MHz)

As with any oscillator, the amplification of the active component should be marginally larger than the attenuation of the capacitive voltage divider, to obtain stable operation. Thus, a Colpitts oscillator used as a variable frequency oscillator (VFO) performs best when a variable inductance is used for tuning, as opposed to tuning one of the two capacitors. If tuning by variable capacitor is needed, it should be done via a third capacitor connected in parallel to the inductor (or in series as in the Clapp oscillator).

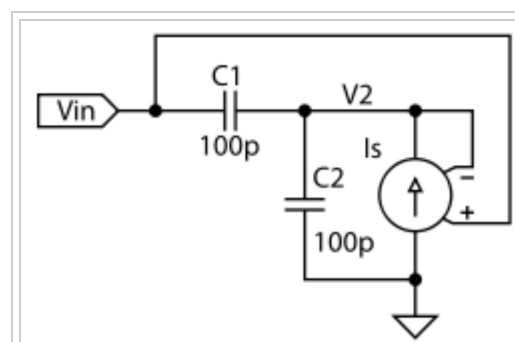
Practical example

Fig. 3 shows a working example with component values. Instead of bipolar junction transistors, other active components such as field effect transistors or vacuum tubes, capable of producing gain at the desired frequency, could be used.

Theory

One method of oscillator analysis is to determine the input impedance of an input port neglecting any reactive components. If the impedance yields a negative resistance term, oscillation is possible. This method will be used here to determine conditions of oscillation and the frequency of oscillation.

An ideal model is shown to the right. This configuration models the common collector circuit in the section above. For initial analysis, parasitic elements and device non-linearities will be ignored. These terms can be included later in a more rigorous analysis. Even with these approximations, acceptable comparison with experimental results is possible.



Colpitts oscillator model used in analysis at left.

Ignoring the inductor, the input impedance can be written as

$$Z_{in} = \frac{v_1}{i_1}$$

Where v_1 is the input voltage and i_1 is the input current. The voltage v_2 is given by

$$v_2 = i_2 Z_2$$

Where Z_2 is the impedance of C_2 . The current flowing into C_2 is i_2 , which is the sum of two currents:

$$i_2 = i_1 + i_s$$

Where i_s is the current supplied by the transistor. i_s is a dependent current source given by

$$i_s = g_m (v_1 - v_2)$$

Where g_m is the transconductance of the transistor. The input current i_1 is given by

$$i_1 = \frac{v_1 - v_2}{Z_1}$$

Where Z_1 is the impedance of C_1 . Solving for v_2 and substituting above yields

$$Z_{in} = Z_1 + Z_2 + g_m Z_1 Z_2$$

The input impedance appears as the two capacitors in series with an interesting term, R_{in} which is proportional to the product of the two impedances:

$$R_{in} = g_m \cdot Z_1 \cdot Z_2$$

If Z_1 and Z_2 are complex and of the same sign, R_{in} will be a negative resistance. If the impedances for Z_1 and Z_2 are substituted, R_{in} is

$$R_{in} = \frac{-g_m}{\omega^2 C_1 C_2}$$

If an inductor is connected to the input, the circuit will oscillate if the magnitude of the negative resistance is greater than the resistance of the inductor and any stray elements. The frequency of oscillation is as given in the previous section.

For the example oscillator above, the emitter current is roughly 1 mA. The transconductance is roughly 40 mS. Given all other values, the input resistance is roughly

$$R_{in} = -30 \Omega$$

This value should be sufficient to overcome any positive resistance in the circuit. By inspection, oscillation is more likely for larger values of transconductance and smaller values of capacitance. A more complicated analysis of the common-base oscillator reveals that a low frequency amplifier voltage gain must be at least four to achieve oscillation.^[6] The low frequency gain is given by:

$$A_v = g_m \cdot R_p \geq 4$$

If the two capacitors are replaced by inductors and magnetic coupling is ignored, the circuit becomes a Hartley oscillator. In that case, the input impedance is the sum of the two inductors and a negative resistance given by:

$$R_{in} = -g_m \omega^2 L_1 L_2$$

In the Hartley circuit, oscillation is more likely for larger values of transconductance and larger values of inductance.

Oscillation amplitude

The amplitude of oscillation is generally difficult to predict, but it can often be accurately estimated

using the describing function method.

For the common-base oscillator in Figure 1, this approach applied to a simplified model predicts an output (collector) voltage amplitude given by:^[7]

$$V_C = 2I_C R_L \frac{C_2}{C_1 + C_2}$$

where I_C is the bias current, and R_L is the load resistance at the collector.

This assumes that the transistor does not saturate, the collector current flows in narrow pulses, and that the output voltage is sinusoidal (low distortion).

This approximate result also applies to oscillators employing different active device, such as MOSFETs and vacuum tubes.

External links

- Java Simulation of a Colpitts oscillator (<http://www.falstad.com/circuit/e-colpitts.html>)

References

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 - [^] Carr, Joe (2002). *RF Components and Circuits* (<http://books.google.com/books?id=V9gBTNvt3zIC&pg=PA127&dq=Colpitts+oscillator+%22capacitive+voltage+divider>). US: Newnes. p. 127. ISBN 0750648449.
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 - [^] Razavi, B. Design of Analog CMOS Integrated Circuits. McGraw-Hill. 2001.
 - [^] Trade-Offs in Analog Circuit Design: The Designer's Companion, Part 1 By Chris Toumazou, George S. Moschytz, Barrie Gilbert [1] (<http://books.google.com/books?id=VoBIOvirkiMC&lpg=PA568&ots=MD4aYrSVjr&dq=the%20tank%20voltage%20amplitude%20is%20calculated%20to%20be&pg=PA568#v=onepage&q=the%20tank%20voltage%20amplitude%20is%20calculated%20to%20be&f=false>)
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